文章编号: 2095-4980(2017)04-0539-04

# 0.14 THz imaging system for security and surveillance

ZHAO Yujiao<sup>a,b</sup>, DENG Xianjin<sup>a,b</sup>, CHENG Binbin<sup>a,b</sup>, LIU Jie<sup>a,b</sup>

(a.Microsystem and Terahertz Research Center, China Academy of Engineering Physics, Chengdu Sichuan 610200, China;
 b.Institute of Electronic Engineering, China Academy of Engineering Physics, Mianyang Sichuan 621999, China)

**Abstract:** Active terahertz-wave imaging systems play a significant role in security and surveillance applications. A 0.14 THz near-field imaging system is described, which consists of a signal generator and acquisition unit, a transceiver front end, a digital signal process unit and a motor control unit. Based on the two-dimensional synthetic aperture technique and the image reconstruction algorithm, this system is capable of producing three-dimensional images of 2 mm lateral resolution and 3 cm range resolution in concealed weapon detection experiments.

Keywords:	terahertz-wave;	personnel surveillance;	reconstruction;	wavenumber	
CLC numbe	<b>r:</b> TN957.52	Document code: A	doi:	10.11805/TKYDA201704.053	9

Conventional security inspection measures are becoming much more difficult to achieve the reliable detection under the requirements of high efficiency and high resolution. The X-ray technology is extensively applied in medical diagnostics and luggage inspection<sup>[1]</sup>, however, it is restricted in personnel surveillance applications based on health considerations. The millimeter-wave and terahertz-wave imaging systems are demanded as the best solutions to the near-field personnel surveillance applications since they are competent in seeing through the common clothing materials without causing destructive ionizing radiations<sup>[2]</sup>.

Much effort has been spent on developing security imaging systems<sup>[3-4]</sup>. Active mm-wave systems<sup>[5-6]</sup> working at frequencies below 80 GHz are already commercially available. The Q-band rotary scanning system<sup>[7]</sup> reduces the measurement time through the rotary scanning method, and the 0.2 THz active imaging system presented by IECAS combines the real aperture and the synthesized aperture<sup>[8]</sup>. These imaging systems show that the higher frequency obtains higher lateral resolution and provides more available bandwidth to obtain higher range resolution. On the other hand, the higher frequency results in a denser spatial sampling, which increases the difficulties in signal processing and image reconstruction<sup>[9]</sup>.

In this paper, a 0.14 THz imaging system is presented based on the active radar and synthetic aperture technique. With the frequency range of 143–148 GHz, this system improves imaging resolution, at the same time, shows good performances in signal processing.

### 1 Signal model and imaging processing

The transmitted stepped-frequency signal can be expressed as

$$s_{t}(t) = \sum_{k=0}^{N-1} e^{j2\pi (f_{0}+k\cdot\Delta f)t} \cdot rect\left(\frac{t-kT-\frac{T}{2}}{T}\right)$$
(1)

Where  $f_0$  is the carrier frequency of the first sub-pulse,  $\Delta f$  is the frequency step, N is the sub-band number, and T is the pulse duration of the sub-pulse.

The echo signal is given as

$$s_{\rm r}(t) = \sum_{k=0}^{N-1} {\rm e}^{{\rm j}2\pi(f_0 + k \cdot \Delta f)\left(t - \frac{2R}{c}\right)} \cdot rect\left(\frac{t - kT - \frac{T}{2} - \frac{2R}{c}}{T}\right)$$
(2)

Accepted date: 2016-10-09; Revised date: 2016-12-15

Where R is the range from the target to the antenna, and c is the speed of light.

The production of the echo signal and the conjugated reference signal can be derived

$$s_{if}(t) = \sum_{k=0}^{N-1} e^{-j2\pi (f_0 + k \cdot \Delta f) \frac{2R}{c}} \cdot rect \left( \frac{t - kT - \frac{T}{2} - \frac{2R}{c}}{T} \right)$$
(3)

With the inverse Fourier Transform to the equation 3, a narrow impulse that relates to the target's range information can be obtained.

While the transmitted signal enhances the range resolution through the large bandwidth, the lateral resolution is improved by the synthetic aperture. The synthetic aperture is performed by scanning the transmitting and receiving antenna over a two-dimensional planar area.

The imaging processing method of the stepped frequency echo signal is the same as the processing method of the linear frequency modulation signal, so the conventional algorithms are still available. In this paper, the three-dimensional image reconstruction of the echo signal is based on the wavenumber domain.

In this imaging system, the transmitting antenna and the receiving antenna are separated, but they are assumed to be at the same location. The approximate location is the midpoint of the two antennas and is assumed to be at the position  $(x', y', z_0)$ . The point target is assumed to be at the position (x, y, z). So the phase shift between the transmitted signal and the echo signal can be expressed as

$$\varphi = 2k\sqrt{(x'-x)^2 + (y'-y)^2 + (z_0 - z)^2}$$
(4)

Where k is the wavenumber that is denoted by  $k = 2\pi f/c$ , and f is the frequency of the transmitted signal.

The reflectivity function of the point target is described by f(x, y, z), so the echo signal received by the transceiver  $(x', y', z_0)$  is

$$s(x',y',k) = \iiint f(x,y,z) e^{-j2k\sqrt{(x'-x)^2 + (y'-y)^2 + (z_0-z)^2}} dxdydz$$
(5)

The Fourier Transform of the equation 5 is given by

$$S(k_x, k_y, k) = \iiint f(x, y, z) \cdot S_1(k_x, k_y, k) dx dy dz$$
(6)

Where  $S_1(k_x, k_y, k) = \int \int e^{-j2kR} e^{-jk_x x' - jk_y y'} dx' dy'$ , and *R* is the range from the target to the antenna which is described by  $R = \sqrt{(x'-x)^2 + (y'-y)^2 + (z_0 - z)^2}$ . Based on the principle of stationary phase, the simplification of  $S_1(k_x, k_y, k)$  is

$$S_{1}(k_{x},k_{y},k) \approx e^{-j\sqrt{4k^{2}-k_{x}^{2}-k_{y}^{2}(z_{0}+z)}}e^{-jk_{x}x-jk_{y}y}$$
(7)

Then the Fourier Transform of the echo signal is

$$S(k_x, k_y, k) = e^{-j\sqrt{4k^2 - k_x^2 - k_y^2 z_0}} \cdot \iiint e^{-jk_x x - jk_y y - jk_z z} f(x, y, z) dx dy dz$$
(8)

Where  $k_z = \sqrt{4k^2 - k_x^2 - k_y^2}$ .

Combining above relations, the reconstruction of the echo signal is

$$s(x, y, z) = IFFT_{(k_x, k_y, k_z)} \left[ S(k_x, k_y, k) \cdot e^{j\sqrt{4k^2 - k_x^2 - k_y^2} z_0} \right]$$
(9)

### 2 System architecture design

The configuration of the 0.14 THz imaging system is shown in Fig.1. This system consists of a signal generator and acquisition unit, a transceiver front end, a digital signal processing unit, and a motor control unit. The signal generator and acquisition unit is utilized to generate a 143–148 GHz stepped-frequency signal and acquire the reflected signal. In this imaging system, the Vector Network Analyzer is adopted to accomplish the signal generator and acquisition. The transceiver front end consists of a power amplifier, a low noise amplifier, and the transceiver antenna. The 143–148 GHz stepped-frequency signal is launched using a wide-beam width antenna, and the reflected signal is received by the receiving antenna. The digital signal processing unit reconstructs the received signal to get a three-dimensional image contains the target information. The motor control unit is employed to control the position of the transceiver antenna.

The base spacing d between the antennas must satisfy the Nyquist criteria in order to obtain a successful discretization. Specifically, the phase shift from one spatial sampling point to the next must be less than  $\pi$  rad. The transmitting antenna beam width and the receiving antenna beam width in



this system are both 60°. Therefore, d is equal to  $\lambda/2$  at 143 GHz (i.e. d=1 mm), where  $\lambda$  is the wavelength that is derived by  $\lambda = c/f$ .

### 3 Results

The 0.14 THz imaging system described in the previous sections is tested in experiments. The transmitted stepped-frequency signal is from 143 GHz to 148 GHz with 201 frequency steps, in which the frequency sampling  $\Delta f$  is 25 MHz. The transmitted power is 1 mW. The spatial separation d is 1 mm. The reflected signal acquired by the Vector Network Analyzer is reconstructed through the equation 5, and the procedure of the reconstruction is shown in Fig.2.



Fig.2 Procedure of the three-dimensional reconstruction

Fig.3 shows an example image of the high quality imaging system. The left side is a photograph of a metal gun at a distance about 0.5 m from the transceiver antennas, while the right side is the reconstruction image. The aperture is  $20 \text{ cm} \times 20 \text{ cm}$  in this experiment. The scanning time now is about 2 h under the restriction of the stepping motor driver, and it can be minimized to several seconds with replacing the mechanical scanning by linear arrays in the further work.



Fig.3 Reconstructed image of a gun

### Reference:

- SHOGO O, ARIMA Y, AKIRA H. Millimeter-wave security imaging using complex-valued self-organizing map for visualization of moving targets[J]. Neurocomputing, 2014(134):247-253.
- SHEEN D M,MCMAKIN D L,HALL T E. Three-dimensional millimeter-wave imaging for concealed weapon detection[J]. IEEE Transactions on Microwave Theory and Techniques, 2001,49(9):1581-1591.
- [3] 成彬彬,李慧萍,安健飞,等. 太赫兹成像技术在站开式安检中的应用[J]. 太赫兹科学与电子信息学报, 2015,13(6):
   843-848. (CHENG Binbin,LI Huiping,AN Jianfei,et al. Application of terahertz imaging in standoff security inspection[J]. Journal of Terahertz Science and Electronic Information Technology, 2015,13(6):843-848.)
- [4] SHEEN D M,FERNANDDES J L,TEDESCHI J R. Wide-bandwidth,wide-beamwidth,high-resolution,millimeter-wave imaging for concealed weapon detection[C]// Proceedings of SPIE-The International Society for Optical Engineering. Baltimore, Maryland, USA:[s.n.], 2013,8715:1-11.
- [5] SHEEN D M,MCMAKIN D L,HALL T E. Near field imaging at microwave and millimeter wave frequencies[C]// IEEE International Microwave Symposium. Honolulu,HI,USA:[s.n.], 2007(9):1693-1696.
- [6] AHMED S S, SCHIESSL A, GUMBMAMN F. Advanced microwave imaging[J]. IEEE Microwave Magazine, 2012, 13(6):26-43.
- [7] GHASR M T,POMMERENKE D,CASE J T,et al. Rapid rotary scanner and portable coherent wideband Q-band transceiver for high-resolution millimeter-wave imaging applications[J]. IEEE Transactions on Instrumentation and Measurement, 2011,60(1):186-197.

- [8] GU S M,LI C,GAO X,et al. Terahertz aperture synthesized imaging with fan-beam scanning for personnel screening[J]. IEEE Transactions on Microwave Theory and Techniques, 2012,60(12):3877-3885.
- [9] 刘玮,李超,张群英,等. 一种用于人体安检的三维稀疏太赫兹快速成像算法[J]. 雷达学报, 2016(3):1-7. (LIU Wei,LI Chao,ZHANG Qunying, et al. Fast three-dimensional sparse holography imaging algorithm for personal security verification[J]. Journal of Radars, 2016(3):1-7.)

### 作者简介:



**ZHAO Yujiao**(1990-), was born in Panzhihua city, Sichuan province. Her current research interests include the area of teraherz radar system and signal processing. email: zhaoyujiao@mtrc.ac.cn.

**LIU Jie**(1981-), was born in Quannan city, Jiangxi province. His current research interests include the area of terahertz imaging system and instruments.

**DENG Xianjin**(1973-), was born in Anyue city, Sichuan province. His current research interests include the area of terahertz communication system.

CHENG Binbin(1982-), was born in Suizhou city, Hubei province. His current research interests include the area of MMW/terahertz radar and imaging system.



# OSEC 2017 首届"兵器工程大会"会议暨征文通知(第一轮)

## The 1<sup>st</sup> Ordnance science and Engineering Conference (OSEC 2017)

兵器工程是一门极具交叉性的工程技术学科,其涉及武器系统与工程、武器发射工程、探测制导与控制技术、 弹药工程与爆炸技术、特种能源技术与工程、装甲车辆工程、信息对抗技术。随着机械、电子、光学、计算机、自 动控制、空气动力学、高温高压流体力学、材料学等学科的发展,不断充实和发展了兵器科学的研究方向,提升了 兵器工程的技术水平,拓宽了兵器工程的应用领域。特别是高新技术的现有成果在兵器工程上的集成,兵器装备与 工程应用已经广泛涉及人工智能、网络信息、动力与能源、近代力学、机械设计与制造、材料与化学、信息光学、 近现代物理等学科的基础理论、技术和方法,并与控制工程、计算机技术、车辆工程、船舶工程、航空航天工程、 电子与通信技术、光学技术等工程领域密切相关。为了更好地引导和推动我国兵器科学与工程的前沿研究及融合发 展,为从事相关领域研究的专家、学者和工程师提供交流平台,拟于 2017 年 10 月在重庆召开首届"兵器工程大会"。

**大会主题: 汇聚交叉与前沿, 推动融合与创新, 促进兵器学科发展。**大会设主论坛"学科交叉与军民融合", 根据论文投稿情况设若干专题进行研讨。如"光机电融合与对抗"专题,"先进动力与推进"专题,"智能制造与装 备"专题,"轻武器设计与制造前沿"专题等。大会将邀请国内外相关领域的权威专家作主题报告和专题报告,以及 优秀论文交流。大会具体内容和征稿具体要求参见 http://scbg.qks.cqut.edu.cn/或 www.iaeej.com, 投稿方式(参会 注册)见 http://bqgcdh.meeting.cos.org.cn, 征文截止时间: 2017 年 9 月 15 日。

# 大会联系方式: 《兵器装备工程学报》编辑部 地址:重庆市巴南区红光大道 69 号 重庆理工大学明德楼 6 楼 614 室 邮 编: 400054 电话: 023-68852703 唐定国、023-62569336 周江川 中国兵工学会 地址:北京市海淀区车道沟 10 号院 中国兵工学会 学术与组织管理部 邮 编: 100089 电话: 15201643738、010-68963055、010-68962962(传真) 葛 萌

十国共工学会 2017-3-8